

Frequency-Domain Measurements

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See <https://mviordache.name/EEGR2051> for more information.

Fourier Series

Any periodic signal $y(t)$ of frequency ω can be represented as a sum of sinusoidal terms of frequency $\omega, 2\omega, 3\omega, \dots$

$$y(t) = c_0 + c_1 \sin(\omega t + \phi_1) + c_2 \sin(2\omega t + \phi_2) + c_3 \sin(3\omega t + \phi_3) + \dots$$

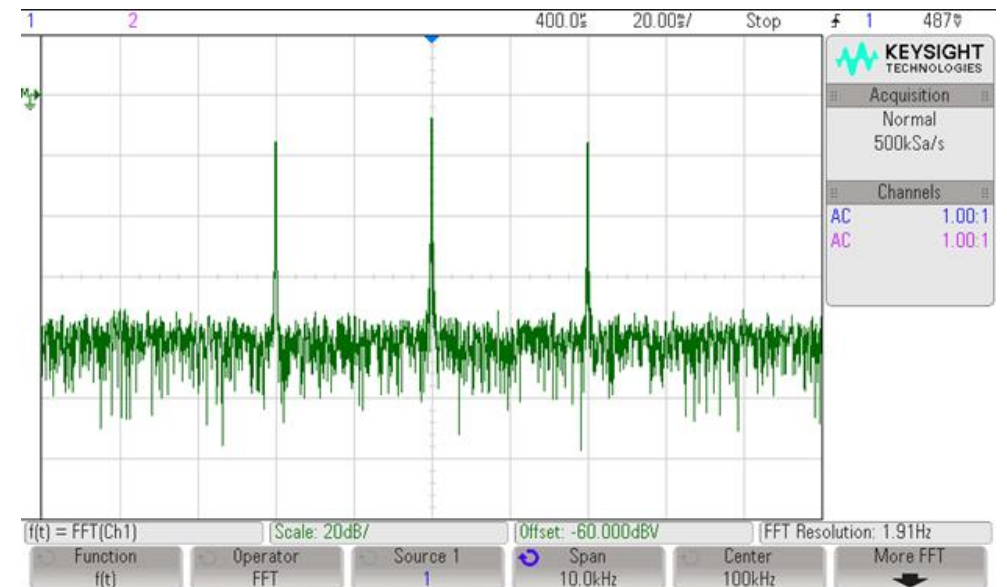
where:

- c_0 is the *DC component* of $y(t)$.
- $c_1 \sin(\omega t + \phi_1)$ is the *fundamental* component of $y(t)$.
- For all $n = 1, 2, 3, \dots$, $c_n \sin(n\omega t + \phi_n)$ is the *harmonic of order n* .

Fourier Series

- Harmonics can be measured with a *spectrum analyzer*.
- Many digital oscilloscopes include an *FFT* option allowing to measure the harmonics of a signal.

Rohde-Schwarz spectrum analyzer. Image [from wikipedia.org](http://from.wikipedia.org).



Fourier Series

- Let $T = \frac{2\pi}{\omega}$ be the period of $y(t)$.
- Harmonics can be calculated with the equations

$$c_0 = \frac{1}{T} \int_0^T y(t) dt, \quad c_n = \sqrt{a_n^2 + b_n^2},$$

$$\cos \phi_n = \frac{b_n}{c_n}, \quad \sin \phi_n = \frac{a_n}{c_n}$$

where

$$a_n = \frac{2}{T} \int_0^T y(t) \cos(n\omega t) dt, \quad b_n = \frac{2}{T} \int_0^T y(t) \sin(n\omega t) dt$$

Example

- The period of the shown waveform is

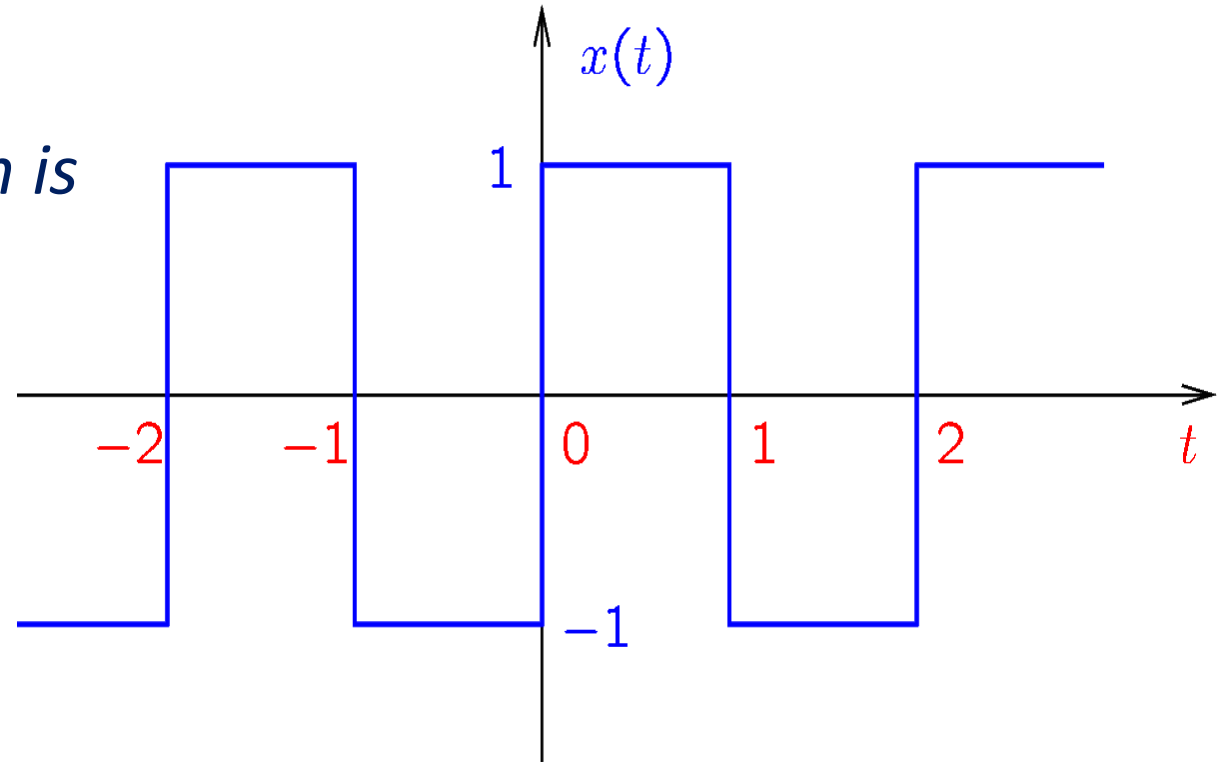
$$T = 2.$$

- The angular frequency of the fundamental is therefore,

$$\omega = \frac{2\pi}{T} = \pi.$$

- It can be shown that

$$x(t) = \frac{4}{\pi} \sin(\omega t) + \frac{4}{3\pi} \sin(3\omega t) + \frac{4}{5\pi} \sin(5\omega t) + \dots$$



Example (continued)

- *Since*

$$x(t) = \frac{4}{\pi} \sin(\omega t) + \frac{4}{3\pi} \sin(3\omega t) + \frac{4}{5\pi} \sin(5\omega t) + \dots$$

- *The average of $x(t)$ is $c_0 = 0$.*
- *The fundamental of $x(t)$ has the amplitude $c_1 = \frac{4}{\pi}$.*
- *The second harmonic of $x(t)$ has the amplitude $c_2 = 0$.*
- *The third harmonic of $x(t)$ has the amplitude $c_3 = \frac{4}{3\pi}$.*

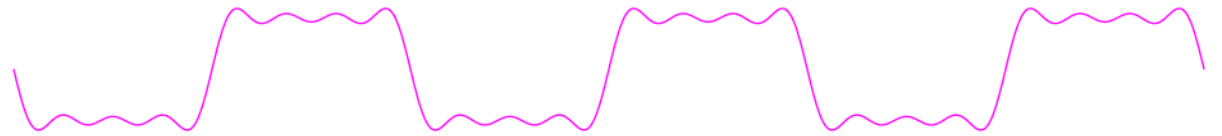
Example (continued)

- The more harmonics are included, the better the approximation!*

$$\frac{4}{\pi} \cos(\omega t) + \frac{4}{3\pi} \cos(3\omega t) \rightarrow$$



$$\frac{4}{\pi} \cos(\omega t) + \dots + \frac{4}{7\pi} \cos(7\omega t) \rightarrow$$



$$\frac{4}{\pi} \cos(\omega t) + \dots + \frac{4}{11\pi} \cos(11\omega t) \rightarrow$$

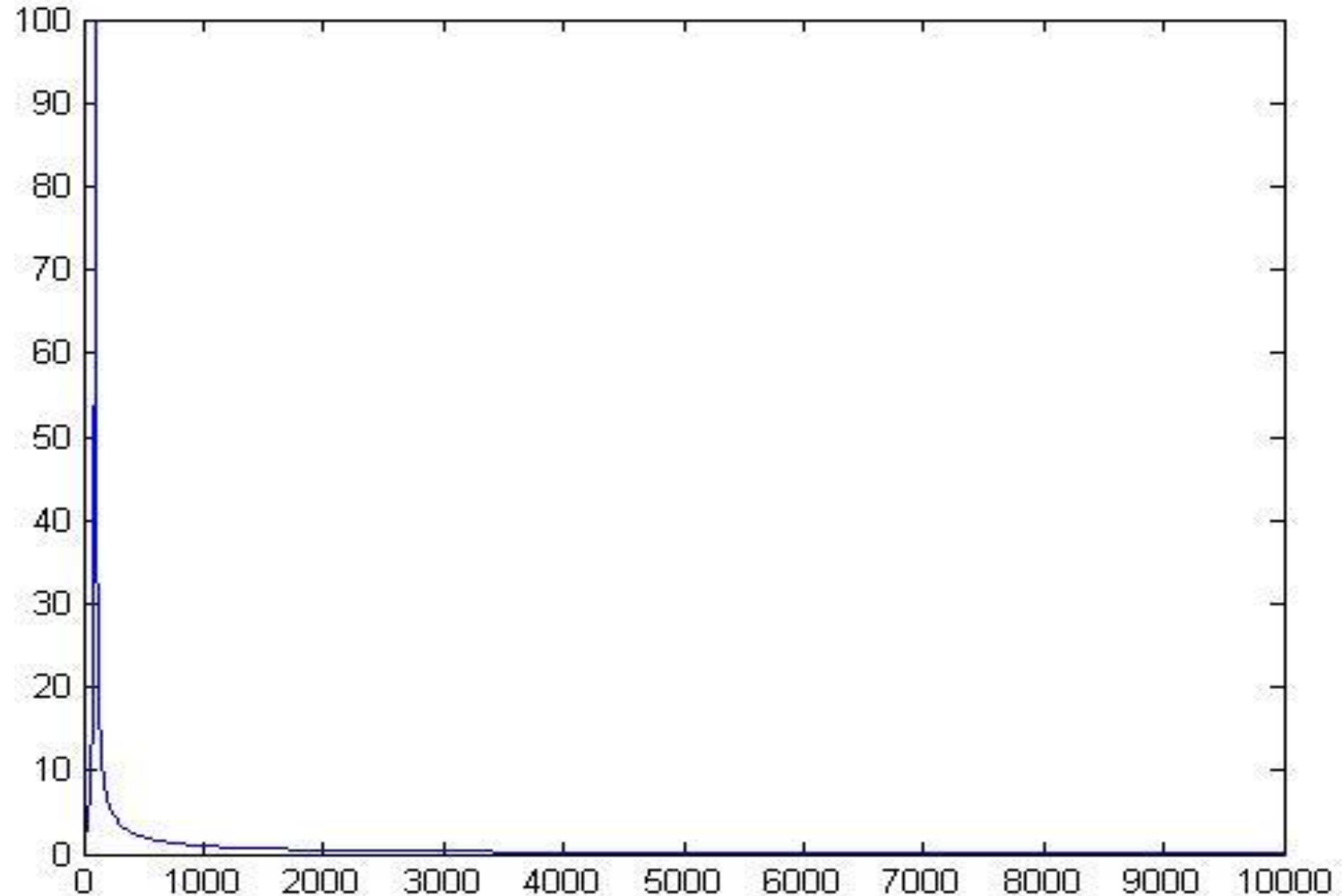


$$\frac{4}{\pi} \cos(\omega t) + \dots + \frac{4}{15\pi} \cos(15\omega t) \rightarrow$$



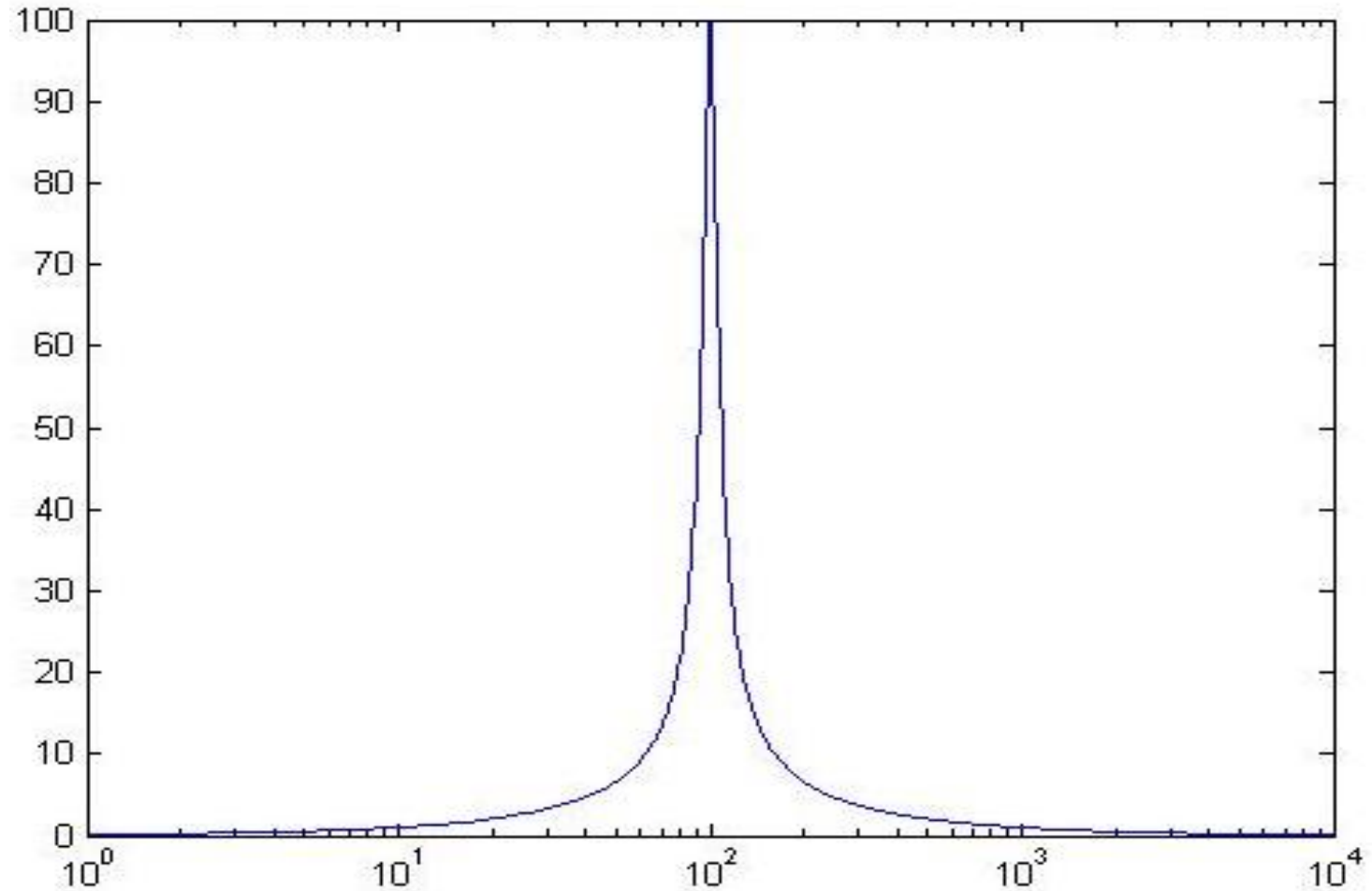
Logarithmic Scale

- The figure shows the graph of Y versus x in blue.
- The values of Y should be read for $x = 10, 100, 1000, \text{ and } 10000$.
- Without question, the values of Y cannot be found precisely by looking at this graph.



Logarithmic Scale

- With a *logarithmic scale* on x , the graph shows Y versus $\log_{10} x$ and labels the x -axis in terms of the values of x (not in terms of $\log_{10} x$.)
- Now the x -axis shows clearly where $x = 10$ and where $x = 100$.
- However, the value of Y at $x = 10000$ cannot be read precisely; all we can tell is that it is close to zero.
- This is because the maximum and minimum values of Y differ by orders of magnitude.

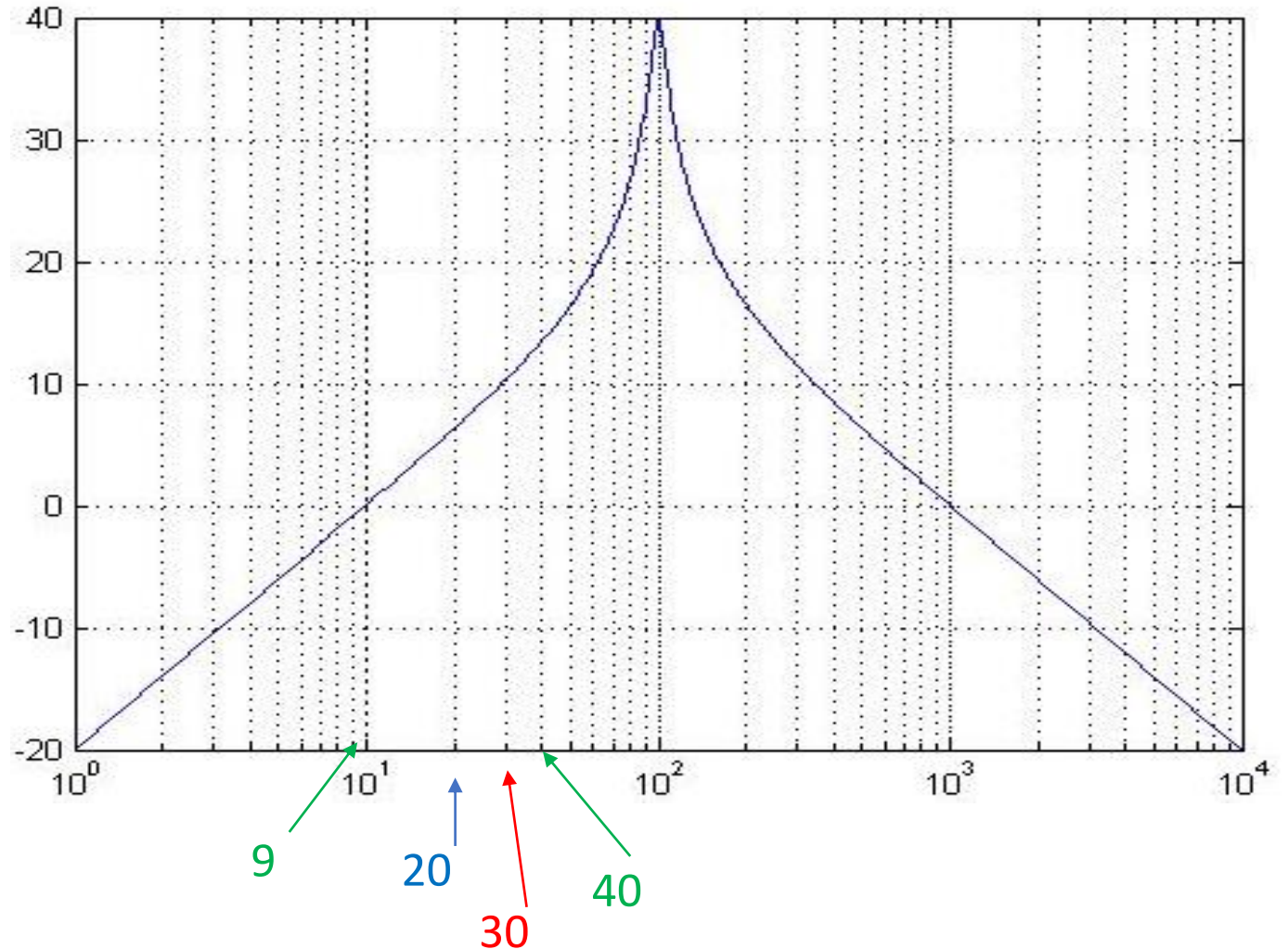


Logarithmic Scale

- The plot in the figure shows Y in decibels versus x , with logarithmic scale on the x -axis.
- With a logarithm on the y -axis, the low values of Y are now clearly seen.
- Note that a voltage Y in volts can be converted to decibels with the formula:

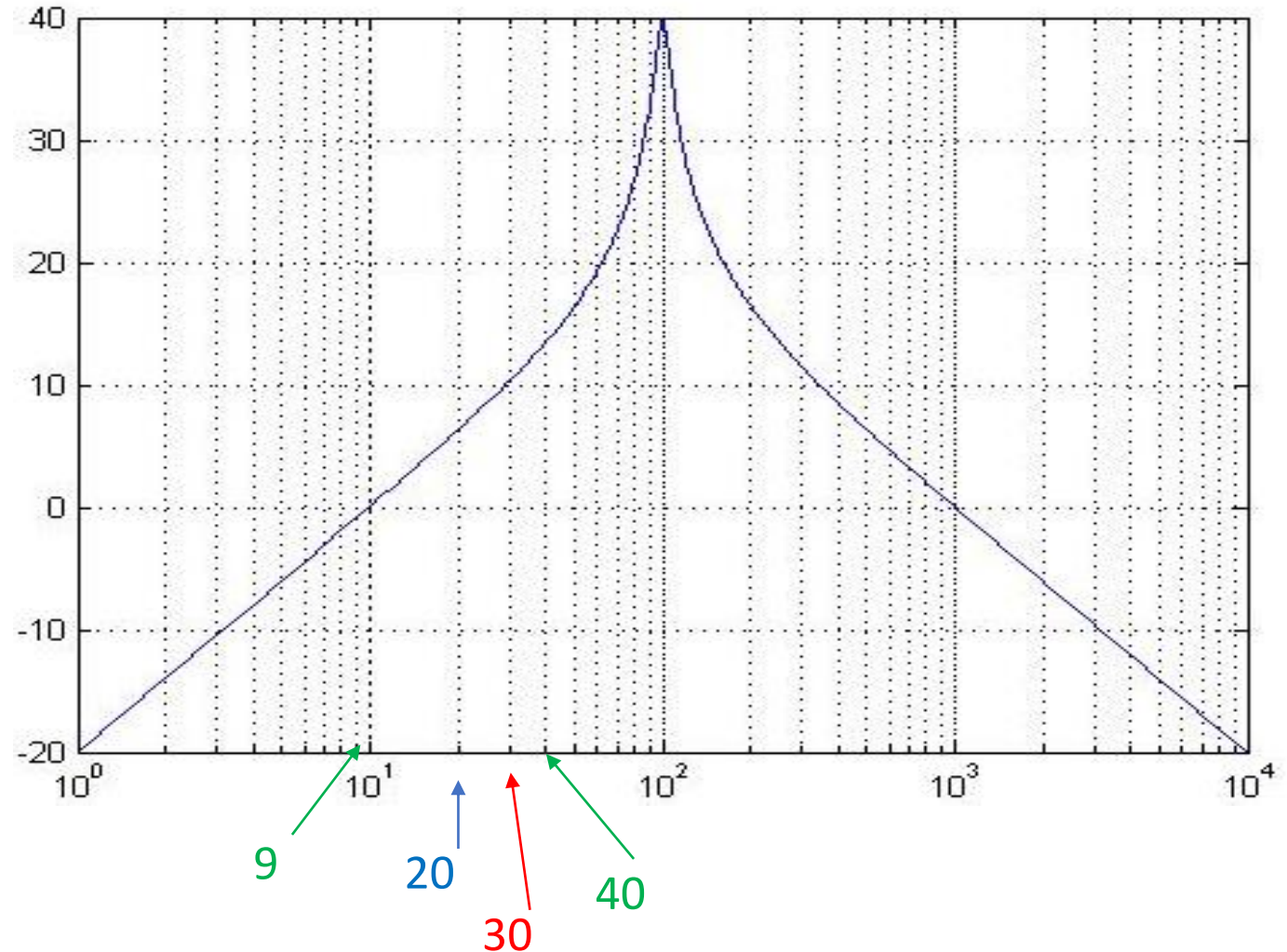
$$Y_{dB} = 20 \log_{10} Y$$

- For example, if $Y = 10 \text{ V}$ rms, then $Y_{dB} = 20 \log_{10} 10 = 20 \text{ dBV}$, where the V suffix of dBV indicates that Y is in volts.



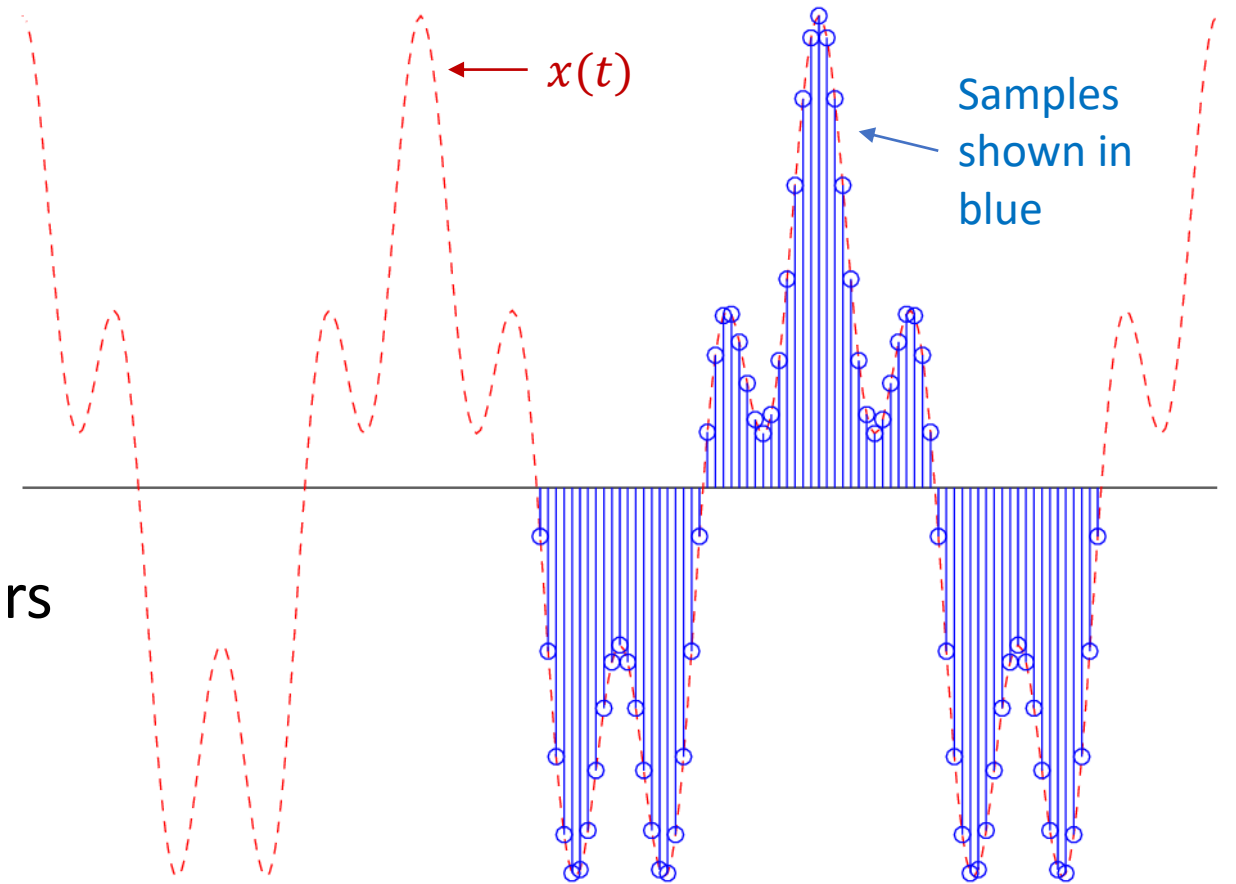
Example

- The vertical gridlines help us read values that are not powers of 10.
- For $x = 10$,
 $Y_{dB} = 0 \text{ dBV} \Rightarrow Y = 1 \text{ V}$.
- For $x = 10^2$,
 $Y_{dB} = 40 \text{ dBV} \Rightarrow Y = 100 \text{ V}$.
- For $x = 10^3$,
 $Y_{dB} = 0 \text{ dBV} \Rightarrow Y = 1 \text{ V}$.
- For $x = 10^4$,
 $Y_{dB} = -20 \text{ dBV} \Rightarrow Y = 0.1 \text{ V}$.
- For $x = 2000$,
 $Y_{dB} = -6 \text{ dBV} \Rightarrow Y = 0.5 \text{ V}$.



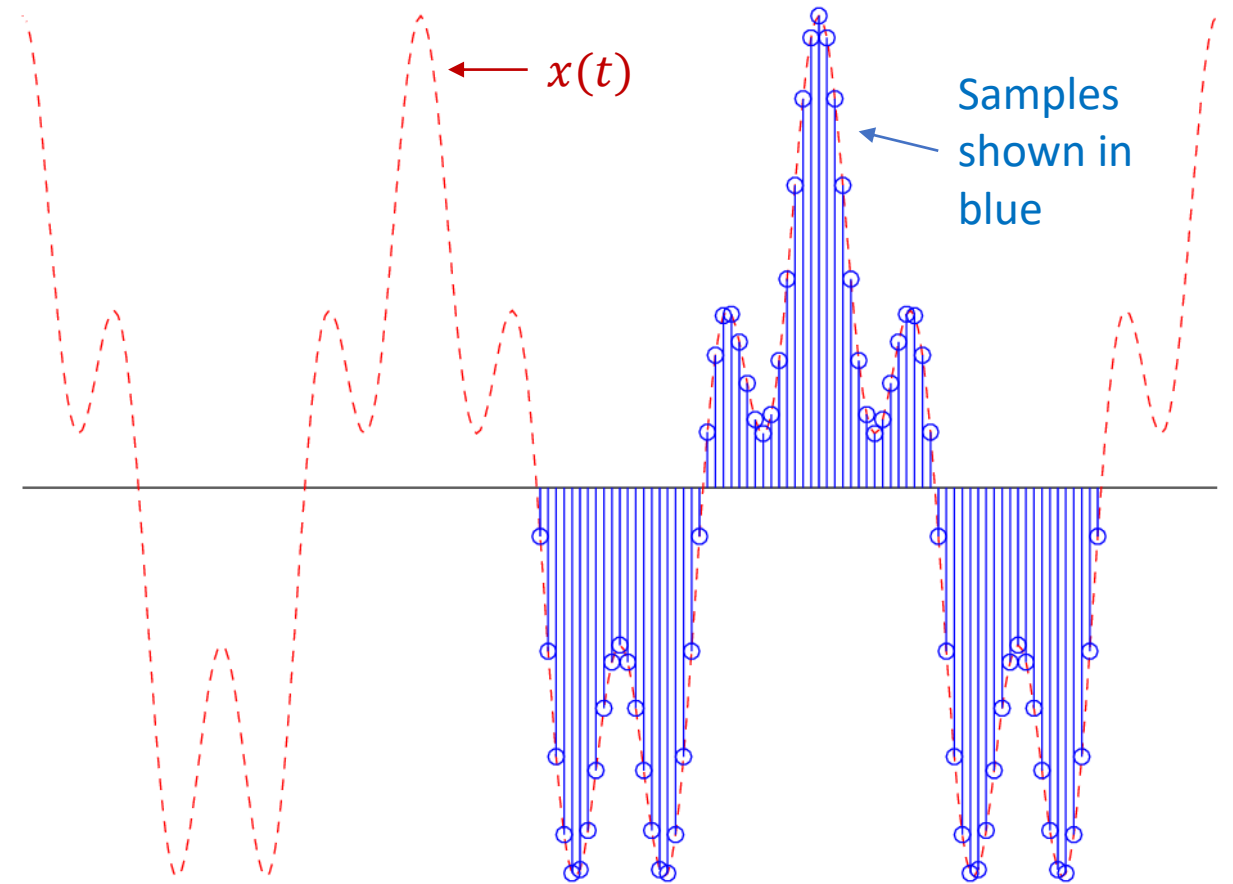
FFT

- Each signal can be decomposed into a sum of sines.
- The *Fast Fourier Transform* (FFT) algorithm provides a way to determine the sines contained into a signal.
- The FFT is applied to N successive samples of a signal $x(t)$.
- Let T be the time interval between two samples.
- The *sampling frequency* is $f_s = \frac{1}{T}$.
- The FFT produces N complex numbers $X(k)$, for $k = 0, 1, \dots, N - 1$.



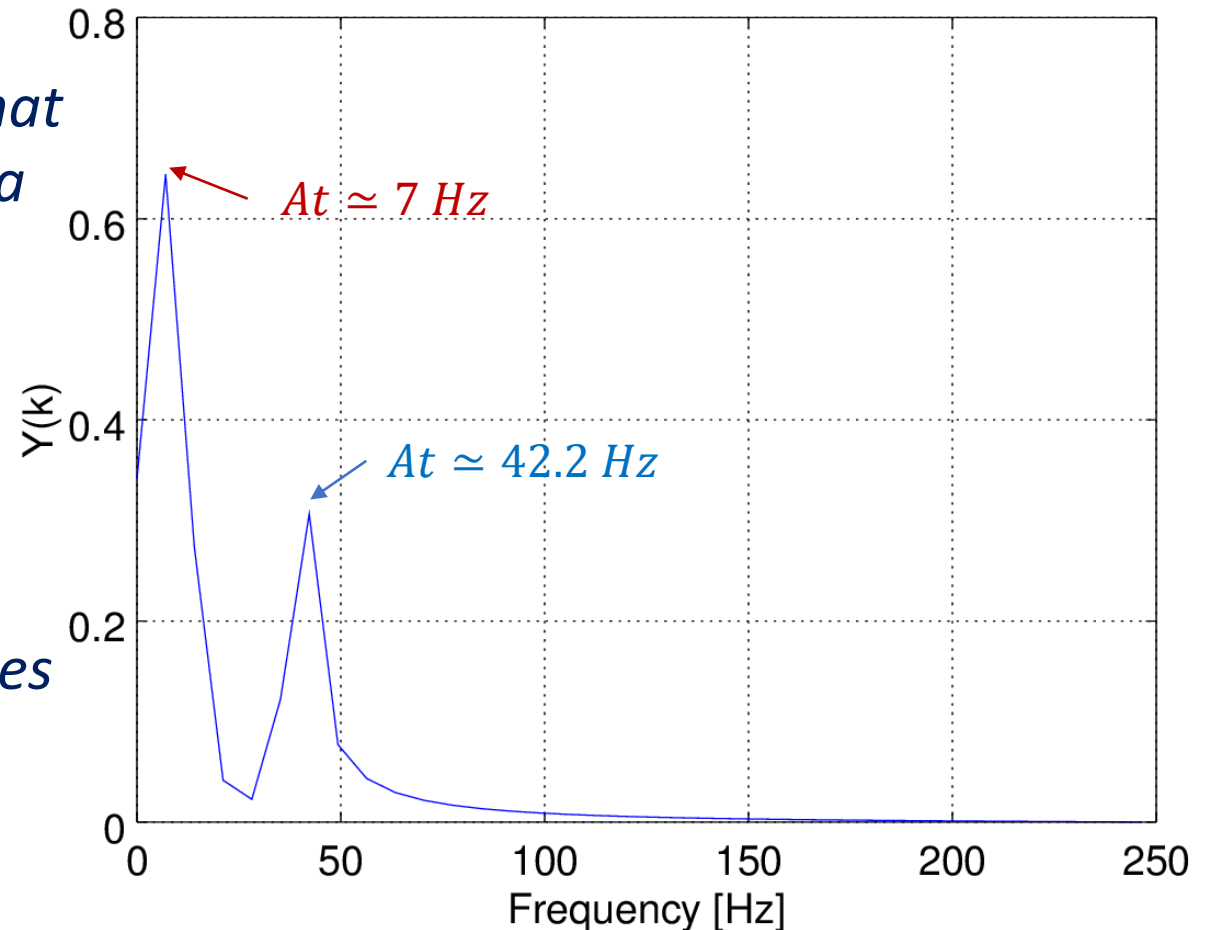
Example

- The figure shows $x(t) = \cos(\omega_1 t) + 0.5\cos(\omega_2 t)$ with $f_1 = 10 \text{ Hz}$ and $f_2 = 40 \text{ Hz}$.
- $N = 71$ samples are taken over a time interval of 142 ms .
- The sampling period is $T = 2 \text{ ms}$.
- The sampling frequency is $f_s = \frac{1}{T} = 500 \text{ Hz}$.



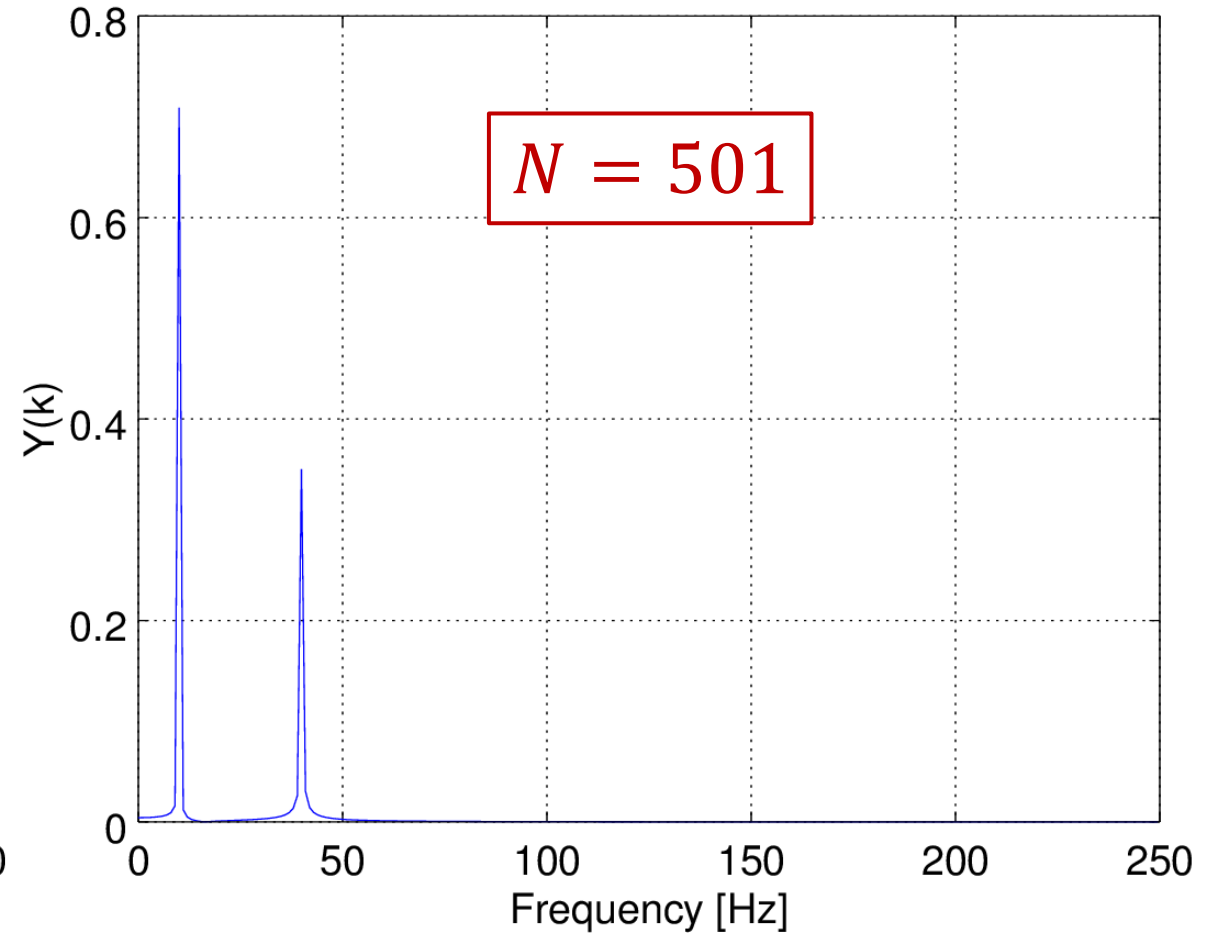
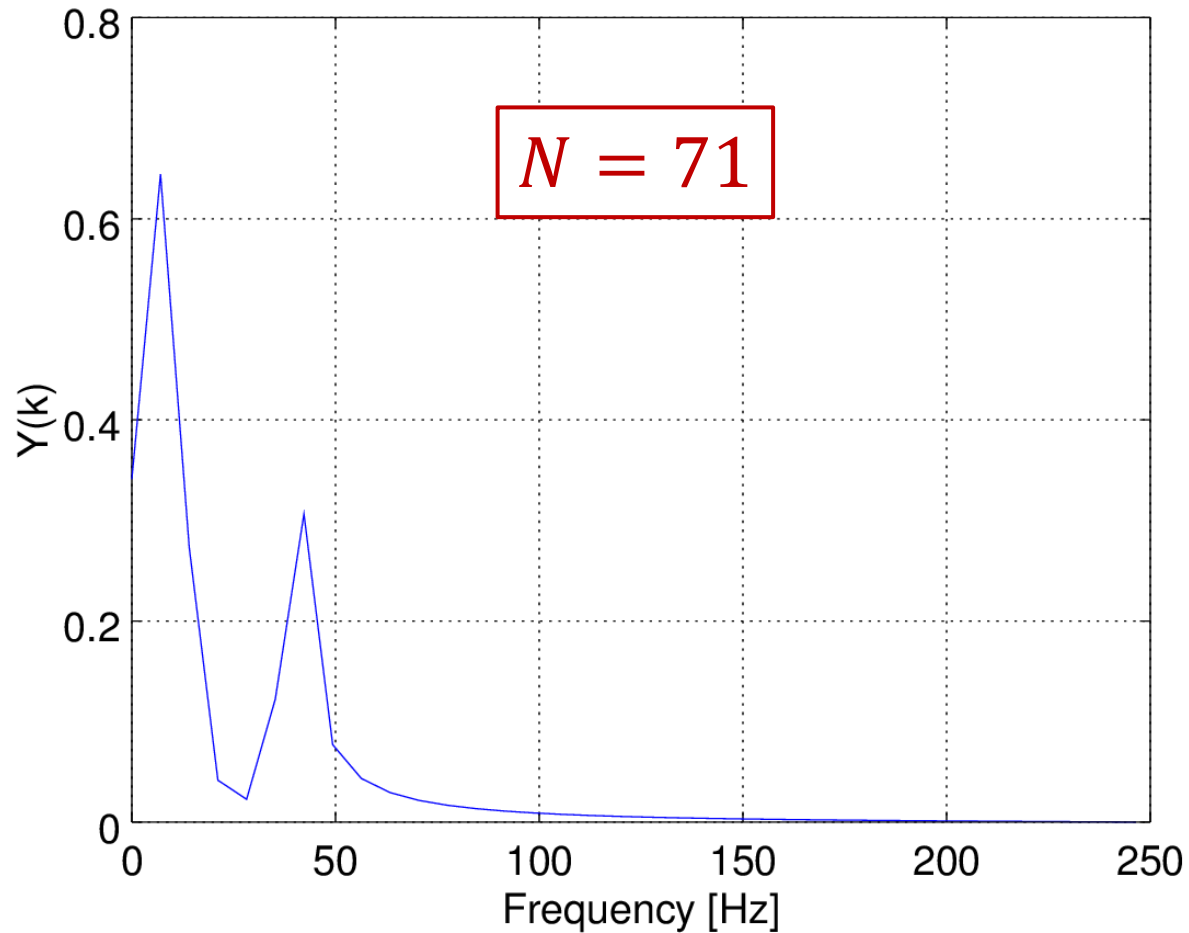
FFT

- The graph $|X(k)|$ vs $k \cdot \frac{f_s}{N}$ for $k = 0, 1, \dots, \left\lfloor \frac{N}{2} \right\rfloor$ reveals the *frequency spectrum* of $x(t)$.
- The horizontal position of the peaks corresponds to the frequencies of the sines that make up $x(t)$.
- *In our example, the FFT graph indicates that $x(t)$ consists of a sine of about 7 Hz and a sine of about 42.2 Hz.*
- The *uncertainty* is $\pm \frac{f_s}{2N} = 3.52$ Hz.
- The *actual signal* is
$$x(t) = \cos(\omega_1 t) + 0.5 \cos(\omega_2 t)$$
with $f_1 = 10$ Hz and $f_2 = 40$ Hz.
- *By using a large enough number of samples N , measurements can be made very precise!*



FFT

- *By using a large enough number of samples N , measurements can be made very precise!*



FFT

- Even with a large number of samples, a small spectral component located close to a large component might not be visible.
- To address this issue, the samples could be multiplied with the weights of a *windowing function*.
 - This reduces the base of the peaks, making it narrower.
- Windowing functions available on lab oscilloscopes include:
 - Rectangular
 - Flat-top
 - Hamming
 - Blackman-Harris
- The *rectangular window* has all weights equal to one.
 - It does not change the samples (it is as if no windowing function were used).

Example

- The graph shows the FFT plot of a signal consisting of three sines.

- Plots are usually scaled so that the vertical axis is in rms volts or dBV.

- Note the zero level.

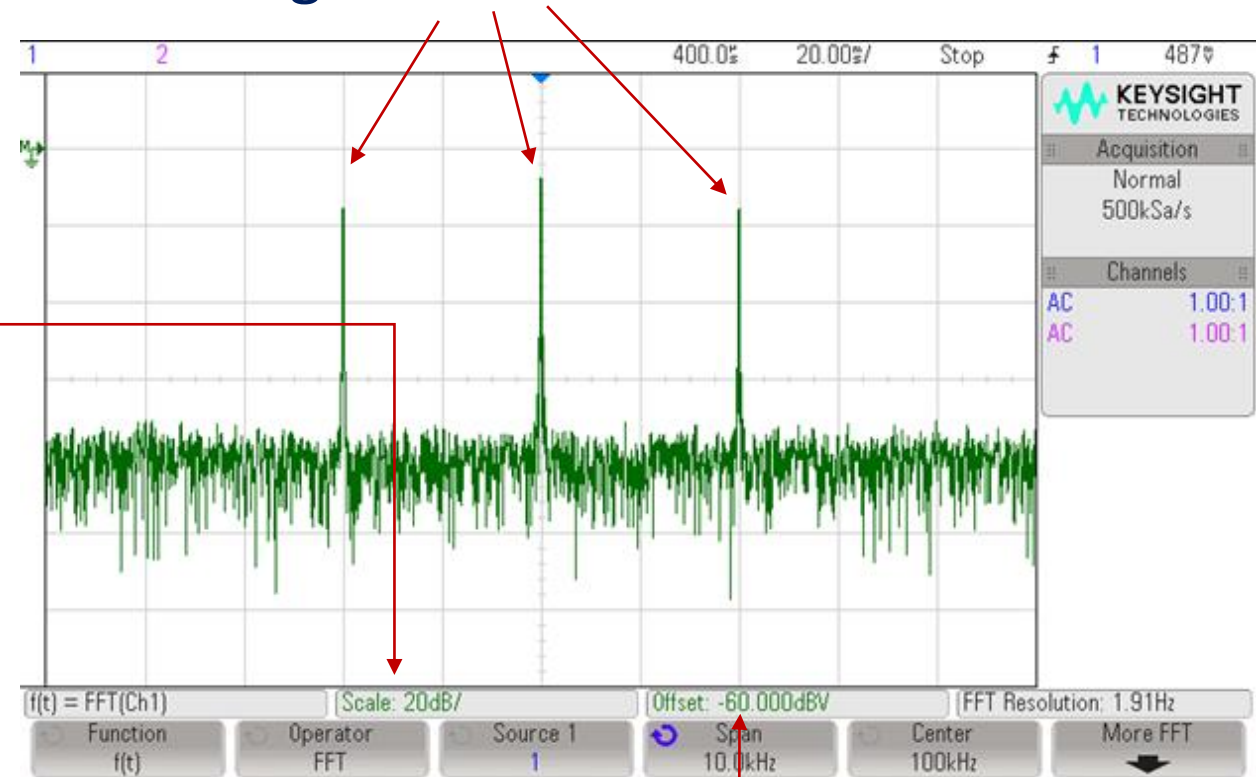
- The vertical scale is in dB.

- The zero level is 3 divisions above the middle of the screen.

- With 20 dB/div , the middle of the screen is at $0 \text{ dB} - 3 \text{ div} \cdot 20 \text{ dB/div}$, that is, at -60 dB .

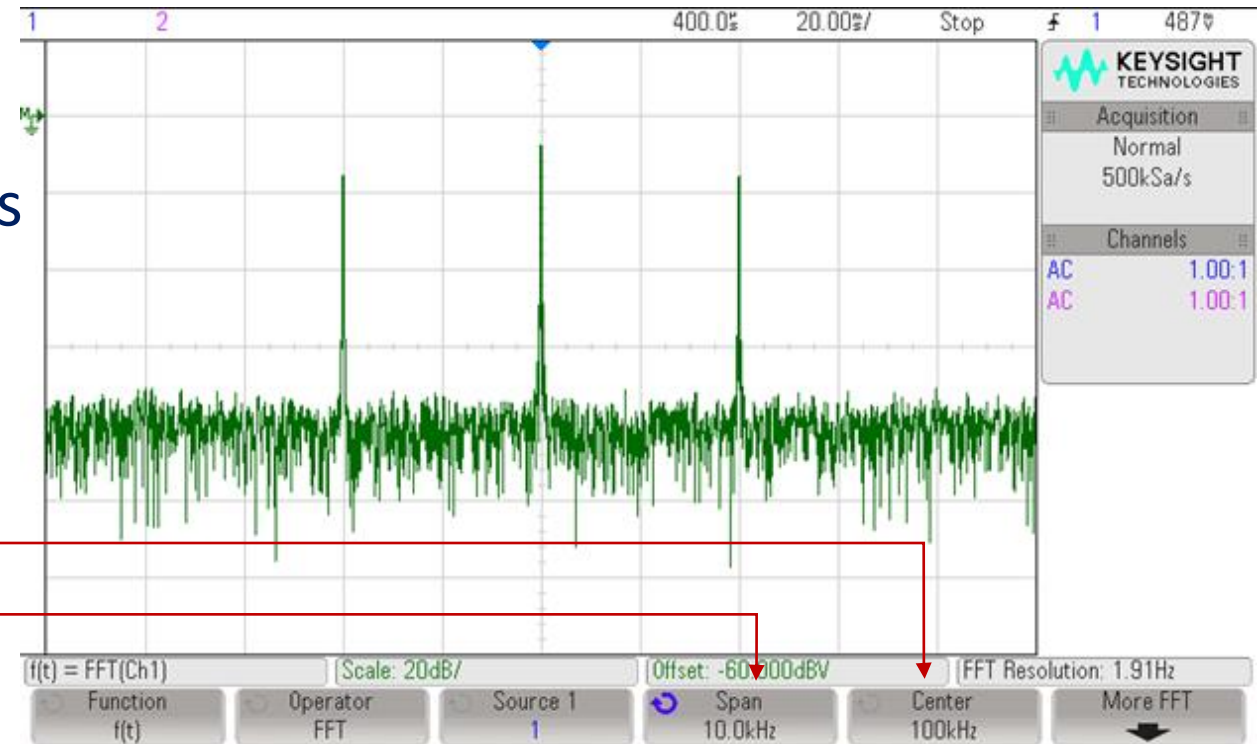
- The rightmost peak is at 0.8 div below the zero level.

- This indicates a sine of $-0.8 \text{ div} \cdot 20 \frac{\text{dB}}{\text{div}} = -16 \text{ dBV}$, that is, 158.5 mV rms .



Example (continued)

- Similarly, the value of the center peak is -0.4 div , that is, -8 dBV , corresponding to 398.1 mV rms .
- On the horizontal axis, the graph has 100 kHz at the center.
- The graph has a span of 10 kHz .
- Since the span has 10 div , we have 1 kHz/div .
- So the center peak is at 100 kHz .
- The right peak is at $100 \text{ kHz} + 2 \text{ div} \cdot 1 \frac{\text{kHz}}{\text{div}} = 102 \text{ kHz}$.
- The left peak is at $100 \text{ kHz} - 2 \text{ div} \cdot 1 \frac{\text{kHz}}{\text{div}} = 98 \text{ kHz}$.



Example (continued)

- We conclude that the graph shows a signal consisting of a sine of 98 kHz, a sine of 100 kHz, and a sine of 102 kHz.
- The amplitudes of the sines can be calculated from the graph.
- The graph does not say anything about the phase angles of the sines.

